Passively Equalized RIAA Phono Preamp using Sonic Imagery Labs Discrete Operational Amplifiers

A current area of high audiophile interest is the passively equalized preamplifier circuits used for phonograph signal sources. The 990Enh-Ticha, 995FET, 992Enh-Ticha and 994Enh DUAL are high performance discrete operational amplifiers designed for professional audio applications and areas where ultra-low noise, low distortion and high output drive current is required. These discrete gain blocks are perfectly suited for phonograph preamp applications.

Introduction
Phonograph preamplifiers differ from other preamplifiers only in their frequency response, which is tailored in a special manner to compensate, or equalize, for the recorded characteristic of a phonograph record. If a fixed amplitude input signal is used to record a phonograph disc, while the frequency of the signal is varied from 20Hz to 20kHz, the playback response curve of Figure 1 will result. Figure 1 shows a plot of a phonograph cartridge output amplitude versus frequency, indicating a severe alteration to the applied fixed amplitude signal. Playback equalization corrects for this alteration and re-creates the applied flat amplitude frequency response. To understand why Figure 1 appears as it does, a detailed explanation of the phonograph recording and cutting process is necessary.

Recording Process and RIAA Curves
The grooves in a stereo phonograph disc are cut by a chisel shaped cutting stylus driven by two vibrating systems arranged at right angles to each other. The cutting stylus vibrates mechanically from side to side in accordance with the signal impressed on the cutter. This is termed a “lateral cut.” The resultant movement of the groove back and forth about its center is known as groove modulation. The amplitude of this modulation cannot exceed a fixed amount or “cutover” occurs. (Cutover, or overmodulation, describes the breaking through the wall of one groove into the wall of the previous groove.) The ratio of the maximum groove signal amplitude possible before cutover, to the effective groove noise amplitude caused by the surface of the disc material, determines the dynamic range of a record (typically 58-60 dB). The latter requirement results from the grainy characteristic of the disc surface acting as a noise generator. (The cutting stylus is heated during master cutting to impart a smooth side wall to minimize the noise.) Of interest in phono preamp design is that the phonograph record noise performance tends to be ten times worse than that of the preamp, with typical wideband levels equal to 10uV of noise or more.

Amplitude and frequency characterize an audio signal. Both must be recorded and recovered accurately for high quality music reproduction. Audio amplitude information translates to groove modulation amplitude, while the frequency of the audio signal appears as the rate of change of the groove modulations. Sounds simple enough, but Figure 1 should, therefore, be a horizontal straight line centered on 0dB, since it represents a fixed amplitude input signal. The trouble results from the characteristics of the phonograph cutting head due to its construction. The cutting head has a resonant peak at 700 Hz and to correct for this resonance, negative feedback coils are added to linearize or flatten the mechanical resonance. Adding the feedback coils produces a velocity output independent of frequency; therefore, the cutting head is known as a constant velocity device.

Figure 1 appears as it does because the cutting amplifier is pre-equalized to provide the recording characteristic shown. Two reasons account for the shape: first, low frequency attenuation prevents cutover and overmodulation of the cutting stylus; second, high frequency boost improves the signal to noise ratio.
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The required RIAA (Record Industry Association of America) playback equalization curve (Figure 2) shows the idealized case dashed line and the actual realization drawn solid. Three frequencies are noted as standard design reference points and are sometimes referred to as time constants. This is a carryover from the practice of specifying corner frequencies by the equivalent RC circuit \( t = RC \) that realized the response. Conversion is done simply with the expression \( t = \frac{1}{2\pi f} \) and results in time constants of 3180us for \( f_1 \), 318us for \( f_2 \), and 75us for \( f_3 \). Frequency \( f_2 \) is referred to as the turnover frequency since this is the point where the system changes from constant amplitude to constant velocity. (Likewise, \( f_3 \) is another turnover frequency.) Table 1 is included as a convenience in checking phono preamp RIAA response.

![Figure 2. RIAA Playback Equalization](chart)

**Figure 2. RIAA Playback Equalization**

Low-Fi Ceramic and Crystal Cartridges

Before getting into the details of building a preamp for magnetic phono cartridges, a few words about crystal and ceramic cartridges are appropriate. In contra-distinction to the constant velocity magnetic pickups, ceramic pickups are constant amplitude devices and therefore do not require equalization, since their output is inherently flat. Referring to Figure 2 indicates that the last sentence is not entirely true. Since the region between \( f_2 \) and \( f_3 \) is constant velocity, the output of a ceramic device will drop 12dB between 500 Hz and 2000 Hz. While this appears to be a serious problem, in reality it is not. This is true due to the inherently poor frequency response of ceramic and restriction of its use to lo-fi and mid-fi market places. Since the output levels of these types of cartridges are so large (100mV-2V), a preamp is not necessary for ceramic pickups; the output is fed directly to the power amplifier via passive tone (if used) and volume controls.

**Table 1. RIAA Standard Response**

<table>
<thead>
<tr>
<th>Hz</th>
<th>dB</th>
<th>Hz</th>
<th>dB</th>
</tr>
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<tbody>
<tr>
<td>20</td>
<td>+19.3</td>
<td>800</td>
<td>+0.7</td>
</tr>
<tr>
<td>30</td>
<td>+18.6</td>
<td>1k</td>
<td>0.0*</td>
</tr>
<tr>
<td>40</td>
<td>+17.8</td>
<td>1.5k</td>
<td>-1.4</td>
</tr>
<tr>
<td>50</td>
<td>+17.0</td>
<td>2k</td>
<td>-2.6</td>
</tr>
<tr>
<td>60</td>
<td>+16.1</td>
<td>3k</td>
<td>-4.8</td>
</tr>
<tr>
<td>80</td>
<td>+14.5</td>
<td>4k</td>
<td>-6.6</td>
</tr>
<tr>
<td>100</td>
<td>+13.1</td>
<td>5k</td>
<td>-8.2</td>
</tr>
<tr>
<td>150</td>
<td>+10.3</td>
<td>6k</td>
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<td>+8.2</td>
<td>8k</td>
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<tr>
<td>400</td>
<td>+3.8</td>
<td>15k</td>
<td>-17.2</td>
</tr>
<tr>
<td>500</td>
<td>+2.6</td>
<td>20k</td>
<td>-19.6</td>
</tr>
</tbody>
</table>

* Reference frequency.

Passively Equalized RIAA Phonograph Preamp Circuit

A circuit topology which can be used for such RIAA phonograph applications is shown in Figure 4. There are many topologies for this application but we have chosen this configuration because it can be easily modified for experimentation. This circuit consists of two high-quality wide-bandwidth-gain blocks, U1 and U2, which are separated by a three-terminal network. The 990Enh-Ticha, 995FET, 992Enh-Ticha and 994Enh DUAL are high performance discrete operational amplifiers designed for professional audio applications and areas where ultra-low noise, low distortion and low offset is required. These discrete gain blocks are perfectly suited for phonograph preamp applications. Each of the gain stages is set up for the required gain, via R3-R4 and R5-R6. Input termination, as appropriate to the particular cartridge used, is provided by Rterm and Cterm, which are optimized for flattest response into this passive network. Since there is no real industry standard for interface, it is important to examine the phono cartridge manufacturers datasheet and understand its output specifications.
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RIAA Equalizer Network
A three-terminal, four-component RC network can be designed to accomplish RIAA equalization, and two examples of such a network are shown in Figure 3. These networks can be used with either the conventional feedback-type of RIAA phono stages (also known as active equalization), or, with the passive equalized preamps between two gain stages.

![RIAA Equalizer Network Diagram](image)

Regardless of the circuit(s) within which they are used, the two RC networks shown are equivalent in terms of achieving the three time constants. Both have an input terminal, a common terminal, and an output terminal. Precision film RC components are recommended for best results. While practical standard values are shown, alternate sets can be attained, simply by multiplying the resistors by a desired factor, while dividing the capacitors by the same factor.

The actual network components should be precision high-quality types, both for initial equalization accuracy and also for minimal errors from nonideal properties. High-quality metal-film resistors and film capacitors of polystyrene or polypropylene are recommended, as they have low-voltage coefficients, low-dissipation factors, and low-dielectric absorption. (A capacitor type which generally should be avoided is the “high K” ceramic families.)

Obviously, the quality of the equalization/amplification can be no better than the components which are used to determine the transfer function (even if the amplifier itself is perfect).

Gain Stages
Refering to schematic of Figure 4. The input RC components, Rterm and Cterm, terminate the moving magnet cartridge, as recommended by the manufacturer. The values shown are typical, with Cterm typically being varied for flattest response. In terms of desired amplifier parameters, for optimum performance in this circuit, they are considerably demanding.

For lowest noise from the cartridge’s inductive source, the amplifier should preferably have a voltage noise density of 5 nV/√Hz or less, and a current noise density of 1 pA/√Hz or less. The former requirement is best met by using low-noise bipolar-input amplifiers, such as the Sonic Imagery Labs 990Enh-Ticha, 992Enh-Ticha or the 994Enh-DUAL. Though the bipolar devices mentioned can meet this requirement for current noise density, it is best met by the use of FET-input amplifiers, such as the Sonic Imagery Labs 995FET-Ticha.

For the desirable low-noise bipolar-input amplifiers, DC input-bias current can be a potential problem when direct coupling to the cartridge; thus, in this circuit, the low-bias-current 995FET-Ticha amplifier is suggested. The amplifier used for U1 should have an input current of less than 100 nA, for minimum problems with the related DC offset (assuming a typical phono cartridge of about 1k resistance). FET-input amplifiers generally have negligible bias currents but they also tend to typically have a higher voltage noise. As mentioned before, the 995FET-Ticha is suitable as it has a reasonably well-controlled noise for this type of application (although not as low as the bipolar devices mentioned).

For high-gain accuracy, particularly at high stage gains, the amplifier should have a high gain-bandwidth product; preferrably 5 MHz or more at audio frequencies.

The op-amp gain blocks, U1 and U2 could be identical for the purposes of simplicity, but are necessarily not so for reasons which are described next.
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The gain values shown yield a 1-kHz gain that is the product of the U1-U2 gains (24.7 times 40.2) times that of the interstage network. For a RIAA-equalized phono case, the 1-kHz gain is 0.101 times the DC gain, which yields an overall gain of 40 dB for this instance. Other gains can be realized most simply by minor increment changes to R4. In general, the 1-kHz gain of this circuit is:

\[ G = 0.101 \left[ 1 + \frac{R3}{R4} \right] \left[ 1 + \frac{R5}{R6} \right] \]

The equalization network which follows U1 should use the lowest impedance values practical from the standpoint of low noise, as the noise output at the network is equivalent to the input-referred noise of U2. As noted previously, scaling can be applied to either network of Figure 3 for component selection, as long as the same ratios are maintained.

A passively equalized preamplifier such as this must be carefully optimized for signal-handling capability, both from an overload standpoint and from a low-noise viewpoint. Stage U1 is desirably chosen for a gain sufficiently high that the input-referred noise will be predominantly due to this stage (and the cartridge, when connected), but yet not so high that it will readily clip at high-level high-frequency inputs. Several amplifiers with a 10-V rms output capability will allow U1 to accept 400-mV rms at high frequencies, as shown (with ±18V supplies). All Sonic Imagery Labs 99X series opamps can operate easily up to ±24V supplies.

Amplifier U2 is less critical than U1 at low frequencies, but, still, is not negligible. A low-noise voltage-density device is very valuable to the U1, and U2 positions, as is a relatively low-input-current noise. As mentioned before, a low bias current is appropriate to U1, with the use of bipolar amplifiers. With a 100nA, or less, bias-current device, direct coupling to a moving magnet phono cartridge is practical. Similarly, the bias-current induced offset voltage of U2, from the 10k DC resistance of the passive RIAA network, will be low relative to the amplified offset of U1. As a result, the worst-case overall-output DC offset at U2 can be held to under 500mV, allowing a single coupling capacitor to suffice for DC blocking purposes for the entire circuit if needed.

In general, the preceding factors dictate that the gain distribution between U1 and U2 be LOW/HIGH from an overload standpoint, but HIGH/LOW from a noise standpoint. Practically, these conflicting requirements can be mitigated by choosing the highest allowable supply voltage for U1 and U2, and the lowest noise device. Because of the near 40dB loss in the passive RIAA network, at 20 kHz, the output overload of the circuit will be noted at high frequencies first. With the gain distribution chosen, the circuit will allow a 3Vrms undistorted output to 20 kHz, with ±15V supplies, and proportionally more with higher supply voltages.
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The low-end rolloff frequency is at the option of the designer. It can be extended towards DC by removing C4, or, alternately, rolled off at a low frequency below 50 Hz. This roll off, when applied, is popularly called a “rumble” filter, and it reduces the turntable/record-induced low-frequency disturbances and lessens the possibility of system and low-frequency driver overloading.

The circuit in Figure 4 can also be easily adapted to servo control any of the output offset by the connection of a noninverting-type servo integrator around stage U2, as shown in Figure 5. The general-purpose noninverting servo described in the Sonic Imagery Labs 990Enh-Ticha opamp datasheet, along with a low-offset FET op amp, such as the AD711 is suitable. Remove C4 if one is to add the servo circuit.

References:


Model 995FET-Ticha Discrete OpAmp
Application Note AN-13

The 995FET-Ticha is a high performance discrete operational amplifier designed for professional audio applications and areas where ultra-low noise and extremely low distortion is required. A true matched monolithic FET input stage is incorporated to provide superior sound quality and speed for exceptional audio performance. This in combination with high output drive capability and excellent dc performance allows use in a wide variety of demanding applications. In addition, the 995FETs wide output swing, allows increased headroom making it ideal for use in any audio circuit.

The 995FET-Ticha can be operated from ±10V to ±24V power supplies. Input cascode circuitry provides excellent common-mode rejection and maintains low input bias current over its wide input voltage range, minimizing distortion. The 995FET discrete op amp is unity-gain stable and provides excellent dynamic behavior over a wide range of load conditions.

The all-discrete design uses an ultra-precision differential matched FET pair specifically designed to meet the requirements of ultra-low noise and ultra-low THD audio systems. In addition to the enhanced input stage, the 995FET uses high performance temperature stable current sources, dual matched pair temperature stable current mirrors and an enhanced low distortion high performance Class-A output driver stage.

Features:
- Ultra Low Total Harmonic Distortion, 0.00055 THD+N @ 1kHz
- Ultra Low Noise, 1.1nV/rtHz
- High Current Output Drive (250mA into 75 ohms)
- +26.5dBu Output Levels (into 600 ohms)
- Standard Gain Block Footprint
- Operates over ±10V to ±24V supply rails
- Lower output offset voltage than existing counterparts
- Lower input leakage current than existing counterparts
- Class A Output Drive
- Particular emphasis on audio performance
- Designed, assembled and produced in the USA
- 3 Year Warranty

Applications:
- High Input Impedance Line Amplifiers and Drivers
- High Input Impedance Buffer
- Active Filters and Equalizers
- Summing/Mixer Amplifiers
- High Performance High Input Impedance Microphone Preamplifiers
- High Performance A/D front end preamplifier
- High Performance D/A back-end driver

Connection Diagram:
Model 990Enh-Ticha Discrete OpAmp
Application Note AN-13

The 990Enh-Ticha is a high performance discrete operational amplifier designed for professional audio applications and areas where ultra-low noise and low distortion is required. It was designed as an enhanced specification upgrade replacement. The pinouts conform to the 990/2520 package, allowing direct replacement. See Table 1. below for additional discrete opamps which can be upgraded. Complete specifications datasheet for the 990Enh-Ticha can be downloaded from www.sonicimagerylabs.com

Table 1. Compatible Upgrade Table
The Model 990Enh-Ticha can be used to upgrade and/or replace these obsolete or end of life discrete operational amplifiers. This list is by no means comprehensive. Contact Sonic Imagery Labs for additional information.

- Jensen JE990 Series
- Automated Processes Inc. API-2520, 2520H, 2525
- John Hardy Co. 990A-990C
- FiveFish Studios DDA series
- Avedis Audio 1122
- Seventh Circle Audio SC10, SC25, SC99
- Sound Skulptor SK25, SK39, SK47
- Yamaha NE80100, NE80200
- TOA PC2011
- ProTech Audio Model 1000
- Purple Audio KDJ3, KDJ4
- Modular Devices 1731, 1757
- Modular Audio Products (MAP) 5000 Series, 1731 1731A
- Melcor 1731
- JLM Audio 99V
- Inward Connections SPA690
- BTI OA400
- FAX Audio FA-100
- Analog Devices 111

Features:
- Ultra Low Total Harmonic Distortion, 0.00045 THD+N @ 1kHz
- Ultra Low Noise <1nV/rtHz (890pV/rtHz typical)
- High Current Output Drive (250mA into 75 ohms)
- +25dBu Output Levels (into 600 ohms)
- Standard Gain Block Footprint
- Operates over ±10V to ±24V supply rails
- Lower output offset voltage than existing counterparts
- Lower input leakage current than existing counterparts
- Particular emphasis on audio performance
- Designed, assembled and produced in the USA
- 3 Year Warranty

Applications:
- Low Impedance Line Amplifiers and Drivers
- Active Filters and Equalizers
- Summing/Mixer Amplifiers
- High Performance Microphone Preamplifiers
- High Performance A/D and D/A front end Preamplifier

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Model 994Enh-Ticha Dual Matched Discrete Operational Amplifier

The 994Enh-Ticha is a dual high performance discrete operational amplifier designed for professional audio applications and areas where ultralow noise and low distortion is required. It was designed as an enhanced upgrade replacement universal dual op-amp gain block. The pinouts conform to the standard 8 pin dual in-line monolithic IC package, allowing direct replacement.

The all-discrete SMT design utilizes an ultra-precision differential matched monolithic transistor pair specifically designed to meet the requirements of ultra-low noise and ultra-low THD audio systems. In addition to the enhanced input stage, the 994Enh-Ticha uses high performance temperature stable current sources, dual matched pair temperature stable current mirrors, dual matched pair active current loads and an enhanced low distortion Class-A output driver stage. Each amplifier is matched for noise, offset and distortion to within 0.1% of each other and both amplifiers meet or exceed published specifications over temperature and operating voltage range.

Because of the 994Enh high current drive capability, supporting circuitry impedances can be scaled down within the application circuit. This can reduce the overall system noise, without increased distortion.

The 994Enh-Ticha op amp is a true bipolar op amp and behaves as such. It does not require a flying ground lead as do other designs on the market. Because the 994Enh is a true op amp, it can also be operated in single supply applications as long as external biasing has been implemented correctly.

Features:
- Ultra Low Total Harmonic Distortion, 0.00045 THD+N @ 1kHz
- Ultra Low Noise 0.89nV/rtHz typical
- High Current Output Drive (100mA into 600 ohms @ ±24V supply)
- +26dBu Output Levels (into 600 ohms @ ±24V supply)
- Standard 8 pin DIP Footprint
- Operates over ±7.5V to ±24V supply rails
- Lower output offset voltage than existing counterparts
- Lower input leakage current than existing counterparts
- Particular emphasis on audio performance
- Designed, assembled and produced in the USA
- 3 Year Warranty

Applications:
- Low Impedance Line Amplifiers and Drivers
- Active Filters and Equalizers
- Summing/Mixer Amplifiers
- High Performance Microphone Preamplifiers
- High Performance A/D and D/A front end Preamplifier
- High Performance D/A I-V convertors
- High Current Buffer Amplifier

Connection Diagram:

Package Diagram:
Model 992Enh-Ticha
Discrete Operational Amplifier

The 992Enh is a high performance discrete operational amplifier designed for professional audio applications and areas where ultralow noise and low distortion is required. It was designed as an enhanced upgrade replacement universal op-amp gain block. The pinouts conform to the standard 8 pin dual in-line monolithic IC package, allowing direct replacement.

The all-discrete SMT design utilizes an ultra-precision differential matched monolithic transistor pair specifically designed to meet the requirements of ultra-low noise and ultra-low THD audio systems. In addition to the enhanced input stage, the 992Enh-Ticha uses high performance temperature stable constant current sources, dual matched pair temperature stable current mirrors, dual matched pair active current loads and an enhanced low distortion Class-A output driver stage.

Because of the 992Enh high current drive capability, supporting circuitry impedances can be scaled down within the application circuit. This can reduce the overall system noise, without increased distortion.

The 992Enh-Ticha op amp is a true bipolar op amp and behaves as such. It does not require a flying ground lead as do other designs on the market. Because the 992Enh is a true op amp, it can also be operated in single supply applications as long as external biasing has been implemented correctly.

Features:
- Ultra Low Total Harmonic Distortion, 0.00045 THD+N @ 1kHz
- Ultra Low Noise 0.89nV/rtHz typical
- High Current Output Drive (100mA into 600 ohms @ ±24V supply)
- +26dBu Output Levels (into 600 ohms @ ±24V supply)
- Standard 8 pin DIP Footprint
- Operates over ±7.5V to ±24V supply rails
- Lower output offset voltage than existing counterparts
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- Particular emphasis on audio performance
- Designed, assembled and produced in the USA
- 3 Year Warranty

Applications:
- Low Impedance Line Amplifiers and Drivers
- Active Filters and Equalizers
- Summing/Mixer Amplifiers
- High Performance Microphone Preamplifiers
- High Performance A/D and D/A front end Preamplifier
- High Performance D/A I-V convertors
- High Current Buffer Amplifier

Connection Diagram:

Package Diagram:

See Also:
Sonic Imagery Labs Model 994Enh-Ticha - Dual Discrete Op Amp DIP8
Sonic Imagery Labs Model 995FET-Ticha - FET Discrete Op Amp 990/2520
Sonic Imagery Labs Model 990Enh-Ticha - Discrete Op Amp 990/2520

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